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Power Allocation for Balancing the Effects of Channel Estimation Error and Pilot Overhead in FD **Decode-and-Forward** Relaying

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Research Questions

- How full-duplex transmission at the relay affects the optimal power allocation between the training and data transmissions?
- How much the system throughput can be improved by using optimal power allocation scheme compared to naive solutions?
- Is low-complexity near-optimal power allocation possible?

System Model

Power Allocation Between Pilots and Data

If $\alpha_x \in (0,1)$ is the fraction of energy devoted to data transmission phase at node $x \in \{S,R\}$ then $C^* = \max_{\alpha_{S}, \alpha_{R} \in (0,1)} \min_{x \in \{SR, RD\}} C_x$ is the achievable rate given optimum power allocation.

- **Problem:** There is *no analytical solution* and *brute-force optimization is very complex*.
- \implies Optimize based on effective SINRs?
- **But:** The SINR of $S \rightarrow R$ channel depends on the estimate \hat{h}_{RR} due to (3); cannot know it *before* allocating powers.
- \implies Replace (3) by a term that does not depend on \hat{h}_{RR} explicitly?

Proposed Algorithm

- 1. Set $|\hat{h}_{RR}|^2 \leftarrow \theta$, where θ is *fixed* parameter.
- 2. Calculate *analytically* optimal $\alpha_{\rm R}^*$ given $\alpha_{\rm S}$.
- 3. Calculate *analytically* optimal $\alpha_{\rm S}^*$.
- 4. If $sinr_{SR} \ge sinr_{RD}$ for assumed SINRs, done!

Optimal *power allocation between training and data* transmission is considered for point-to-point links in [10, 11]. Here similar optimization problem is considered for the *decode-and-forward fullduplex relay channel* when the relay suffers from self-interference due to hardware impairments.



The received *signal at the relay* reads (here $x_{\rm R}$ is known)

 $y_{\rm R} = h_{\rm SR} x_{\rm S} + h_{\rm RR} (x_{\rm R} + m_{\rm R}) + n_{\rm R}$

$$\xrightarrow{-h_{\mathrm{RR}}x_{\mathrm{R}}} y_{\mathrm{R}} = \hat{h}_{\mathrm{SR}}x_{\mathrm{S}} + \left[\Delta h_{\mathrm{SR}}x_{\mathrm{S}} + \hat{h}_{\mathrm{RR}}m_{\mathrm{R}} + \Delta h_{\mathrm{RR}}(x_{\mathrm{R}} + m_{\mathrm{R}}) + n_{\mathrm{R}}\right], \quad (1)$$

where $\hat{h} = h - \Delta h$ is the channel estimate and $\Delta h \sim CN(0, \Delta g)$ is the estimation error. We also let $n_{\rm R} \sim {\rm CN}(0,1)$ and $m_{\rm R} \sim {\rm CN}(0,\sigma_m^2)$.

- **Problem:** "Noise" in (1) depends on x_S and is not Gaussian.
- Solution: Consider a modified $S \rightarrow R$ channel model

$$y_{\rm R} = \hat{h}_{\rm SR} x_{\rm S} + w_{\rm R}$$

(2)

where $w_{\rm R} \sim CN(0, \sigma_{w_{\rm R}}^2)$ is *independent of* $x_{\rm S}$ with

$$\sigma_{w_{\mathrm{R}}}^{2} = 1 + P_{\mathrm{S}}^{\mathrm{d}} \Delta g_{\mathrm{SR}} + P_{\mathrm{R}}^{\mathrm{d}} \Delta g_{\mathrm{RR}} + (|\hat{h}_{\mathrm{RR}}|^{2} + \Delta g_{\mathrm{RR}})\sigma_{m}^{2}$$
(3)

and $P_{\rm S}^{\rm d}$ (resp. $P_{\rm R}^{\rm d}$) is data symbol power at source (resp. relay). The resulting *ergodic link-rate* for $S \rightarrow R$ (similarly for $R \rightarrow D$) is

$$C_{\rm SR} \propto \mathbf{E} \left\{ \log \left(1 + \frac{P_{\rm S}^{\rm d} |\hat{\boldsymbol{h}}_{\rm SR}|^2}{\sigma_{w_{\rm R}}^2} \right) \right\}$$
(4)

where the expectation is w.r.t. $(\hat{h}_{SR}, \hat{h}_{RR})$. By (3), the rate of $S \rightarrow R$ link depends on the power allocations both at the source and relay.

• The achievable rate (4) is a lower bound to true capacity of (1).

- But: The algorithm may not converge to optimal power allocation anymore.
- 5. **Else** solve $sinr_{SR}(\alpha_R) sinr_{RD}(\alpha_R) = 0$ numerically, given analytically optimized α_{s}^{*} .

Proposed algorithm always converges to a solution and does not need calculation of expectations.

Numerical Examples

For the examples below, we set $\sigma_m^2 = 10^{-3}$ and let h_{SR} , $h_{\text{RD}} \sim \text{CN}(0, 1)$ along with $h_{\text{RR}} \sim \text{CN}(0, g_{\text{RR}})$. Both of the examples below also use $\theta_{avg} = E\{|\hat{h}_{RR}|^2\} = g_{RR} - \Delta g_{RR}$ scaled (ad-hoc) by 1/2 for the θ .



Achievable rate vs. the average per-symbol power P_{avg} . Self-interference channel strength is set to $g_{RR} = 30$ dB. Solid lines = proposed algorithm, markers = brute force.

Achievable rate vs. *self-interference channel power* g_{RR} . The two sets of curves correspond to average symbol powers $P_{\text{avg}} \in \{0, 10\} \text{ dB}.$

Conclusions

Power allocation between pilots and data in FD decode-and-forward relay channel was studied.

- A modified channel model that allowed achievable rate analysis was developed.
- Optimal power allocation was found to improve the achievable rates up to 1 bits/s/Hz.
- Proposed low-complexity power allocation scheme is near-optimal for all considered cases.
- Algorithm details and extension to Ricean fading self-interference channel are in the paper.

References

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