

Achievable Transmission Rates and Self-Interference Channel Estimation in Hybrid Full-Duplex/Half-Duplex MIMO Relaying

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- Recent findings, e.g.¹, have clearly shown that inband full-duplex is technically feasible
 - Here we focus on full-duplex relay applications
- The next question is how to best use the fullduplex capability
 - Always transmitting and receiving at the same time, or perhaps choosing the operation mode more wisely?

¹ M. Heino, D. Korpi, T. Huusari, E. Antonio-Rodríguez, S. Venkatasubramanian, T. Riihonen, L. Anttila, C. Icheln, K. Haneda, R. Wichman, and M. Valkama, "Recent Advances in Antenna Design and Interference Cancellation Algorithms for In-band Full-Duplex Relays," *IEEE Communications Magazine*, vol. 53, no. 5, pp. 91-101, May 2015.



- In this work, we explore this question in a simple two-hop relay system with a full-duplex capable relay (R)
 - What kind of transmission scheme eventually <u>maximizes the end-to-end rate</u> between the source (S) and the destination (D) ?





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Or some combination of the two ?

- Even though pure full-duplex relaying might be conceptually the most appealing option, many practical issues affect end-to-end performance
- For instance, the SR and RD links might have different path losses, resulting in different SNRs and thus different achievable data rates between the two links
- Also, for accurate self-interference (SI) cancellation, the SI channel must be estimated at some point
 - This is challenging to do if the relay is also receiving a signal, since the useful RX signal acts as noise from SI channel estimation perspective
 - i.e., there are tradeoffs

Proposed scheme

- In this work, we actually propose a scheme that combines the half-duplex and full-duplex approaches
 - The interesting question is whether it performs better than either the pure full-duplex or pure half-duplex scheme
- This can be determined by evaluating the relay system end-to-end performance in different scenarios
 - <u>Taking into account the inherent transmission capability</u> as well as the performance limitations due to SI channel estimation errors and RF component imperfections

Proposed scheme



- The lengths of the different communication modes can be adjusted
- Next we seek to understand what is the optimum partitioning between HD/FD modes, and how does that link to SI channel estimation



Achievable source-todestination throughput

• The S-D throughput of the system can be expressed as follows:

$$C = \min\left\{ \left(\frac{\tau_{SR}}{\tau_{tot}}\right) C_{SR}^{HD} + \left(\frac{T_{coh} - \tau_{RD}}{\tau_{tot}}\right) C_{SR}^{FD}, \left(\frac{\tau_{RD}}{\tau_{tot}}\right) C_{RD}^{HD} + \left(\frac{T_{coh} - \tau_{RD}}{\tau_{tot}}\right) C_{RD}^{FD} \right\}$$
Source-to-relay
Relay-to-destination

- τ_{SR} , and τ_{RD} , are the times spent in HD mode, T_{coh} is the SI channel coherence time and $\tau_{tot} = T_{coh} + \tau_{SR}$
- The time ratios determine the proportion of time spent in each mode during the relaying procedure
- The <u>achievable end-to-end rate</u> is determined by the smaller of these two rates (S-R, R-D)



Achievable source-todestination throughput

• The <u>rate expressions</u> are here calculated using the well-known Shannon-Hartley theorem, i.e.,

$$C_x^{\mathcal{Y}} = \sum_{i=1}^{N_x} \log_2(1 + \operatorname{sinr}_{i,x}^{\mathcal{Y}} \lambda_{i,x}^2)$$

- Here, $x \in \{SR, RD\}$, $y \in \{HD, FD\}$, N_x is the number of spatial streams, $sinr_{i,x}^{\mathcal{Y}}$ is the signal-to(-interference-plus)-noise ratio, and $\lambda_{i,x}$ is the *i*th singular value of the channel matrix
- The <u>SINR is defined by the received signal power, noise power</u> and in the FD relay also by the SI cancellation performance
 - This is greatly affected by the presence of the source-to-relay transmission during SI channel estimation
 - More detailed SINR expressions available in the paper

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Numerical evaluations



Parameter	Value	Parameter (cont.)	Value
Signal bandwidth	12.5 MHz	Total transmit power of the relay	20 dBm
Waveform	OFDM	Analog SI attenuation at the relay	70 dB
Number of TX antennas at the source	2	IRR at relay (TX & RX)	25 dB
Number of RX/TX antennas at the relay	2/2	Relay TX PA IIP3	18 dBm
Number of RX antennas at the destination	2	Relay RX ADC bits	12
SNR per receiver at the relay, SNR_R	15 dB	SI channel estimation sample size at the relay	5000
SNR per receiver at the destination, SNR _D	20 dB	SI channel coherence time	1 ms

Widely linear digital canceller is adopted at the relay to cancel the SI in the digital domain
 Can process the IQ mismatch but not PA induced nonlinear distortion

Side-step: Reference RF measurements



PSD at different cancellation stages, transmit power +20 dBm



- These measurements were done using a novel relay antenna design and a self-adaptive RF canceller
- Approximately 75 dB of total analog SI cancellation was achieved
- The measurements were done with an LTE waveform at 2.47 GHz and a transmit power of +20 dBm





Here SNR at relay, SNR_R , fixed at 15 dB

- Throughputs for the *proposed scheme* with different RX signal
 strengths at the
 destination
- The length of the HD
 RD transmission
 period is varied
- Throughput is maximized when the SI channel can be estimated without interference
- SI channel estimation sample size is fixed (equal to vertical dashed line)

- Next, five different communication schemes are compared:
 - *Flexible hybrid scheme*, which has no limitations on the lengths of the communication periods (as long as $\tau_{RD} \leq T_{coh}$)
 - Fixed hybrid scheme, where the value of τ_{RD} is chosen such that it corresponds to the time required for estimating the SI channel
 - Pure full-duplex scheme, where $\tau_{SR} = \tau_{RD} = 0$
 - Pure half-duplex scheme, where the transmission periods τ_{SR} and τ_{RD} are chosen optimally to maximize the relay-todestination throughput
 - Pure half-duplex scheme with equal transmission periods, where $\tau_{SR} = \tau_{RD}$





- The flexible and fixed hybrid schemes perform best, thanks to interference-free SI channel estimation
 - It is optimal to choose the length of the HD RD period to match the length of the SI channel estimation period
- With low RX signal power at the relay, also a half-duplex period between source and relay is needed if exactly optimum solution is targeted





- The transmit power has a direct effect on the SINR at the relay due to RF impairments
- With the chosen
 simulation parameters,
 20 dBm is the highest
 feasible transmit
 power
- After this, the
 throughput of the SR
 link drops too low
 because widely linear
 digital canceller cannot
 anymore suppress all
 remaining SI 15

Conclusion

- Both half- and full-duplex communication periods are needed to maximize the throughput of a two-hop link
- In particular, the full-duplex relay should estimate its SI channel during a half-duplex period
- These findings help in optimizing the deployment of full-duplex MIMO relays in future mobile networks



Thank you !

Questions ?

