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School of Science  
and Technology

# Residual Self-Interference in Full-Duplex MIMO Relays After Null-Space Projection and Cancellation

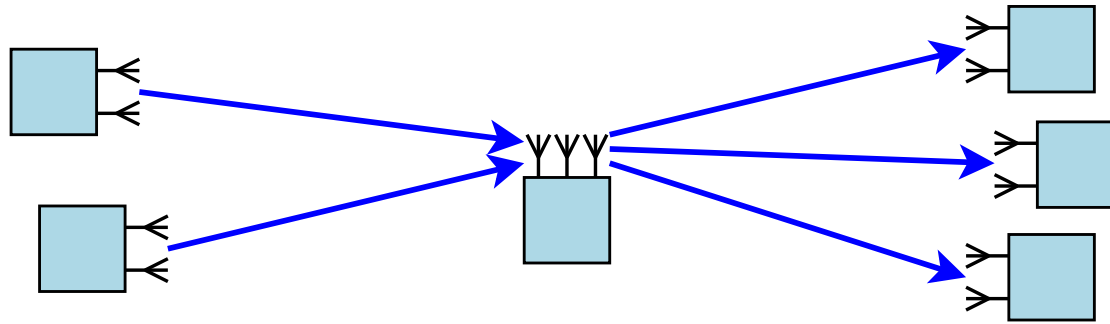
Taneli Riihonen, Stefan Werner, and Risto Wichman  
Aalto University School of Science and Technology

Session MP2b “MIMO Relays”, November 8, 2010  
44th Asilomar Conference on Signals, Systems, and Computers

# Introduction

# Background

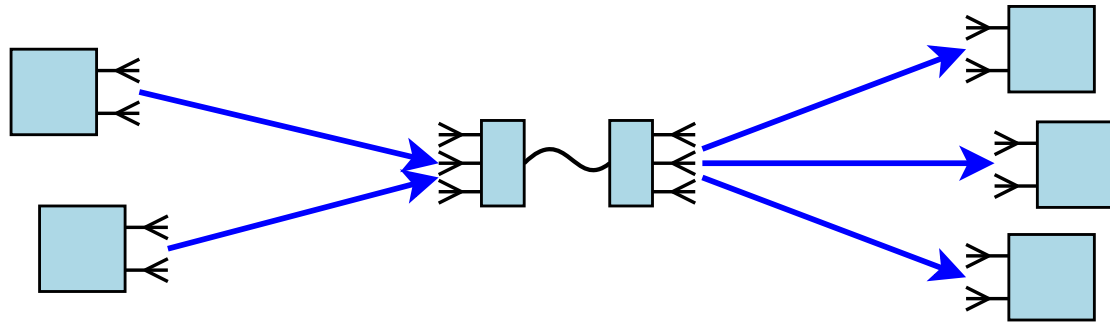
- Two-hop MIMO relay links have been a hot research topic recently



- Most of the literature assumes half-duplex relaying mode
  - ▷ Different time slots or frequency bands for relay rx and tx
  - ▷ **Disadvantage:** Loss in spectral efficiency
  - ▷ Inevitable choice for relays with single antenna array: Weaker desired signal will be drowned by strong self-interference

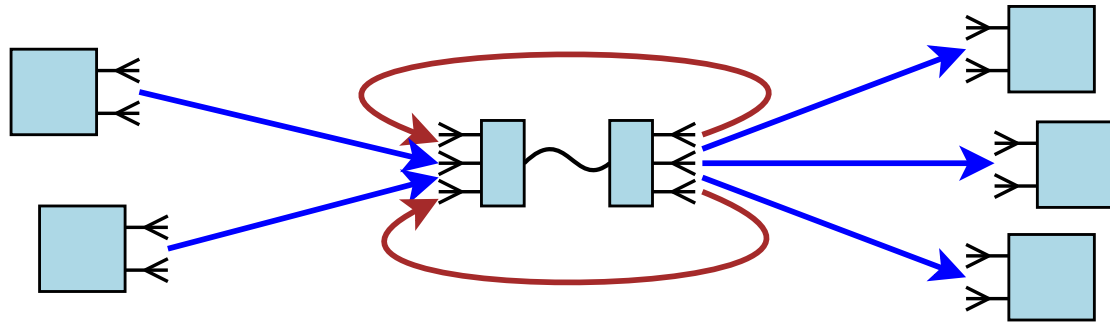
# Goal: Improved Spectral Efficiency

- MIMO relay with separated receive and transmit antenna arrays



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- MIMO relay with separated receive and transmit antenna arrays

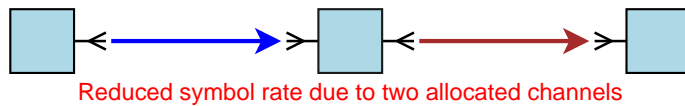


- Now it is viable to choose the full-duplex mode, and avoid the loss of spectral efficiency which is inherent for the half-duplex mode
  - ▷ Natural isolation facilitates the usage of signal processing techniques for mitigating the self-interference
- **Practical problem:** There will be still residual self-interference

# Motivation from Our SISO Relay Studies (1)

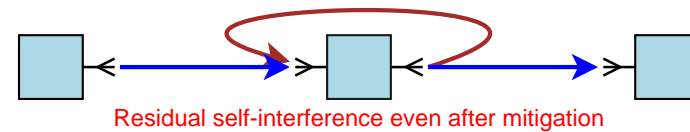
- The choice between half-duplex and full-duplex relaying represents a fundamental rate–interference tradeoff

▷ Half-duplex relay link:



$$\mathcal{R}_{\text{HD}} = \frac{1}{2} \log_2 \left( 1 + \frac{P_S}{P_N} \right)$$

▷ Full-duplex relay link:

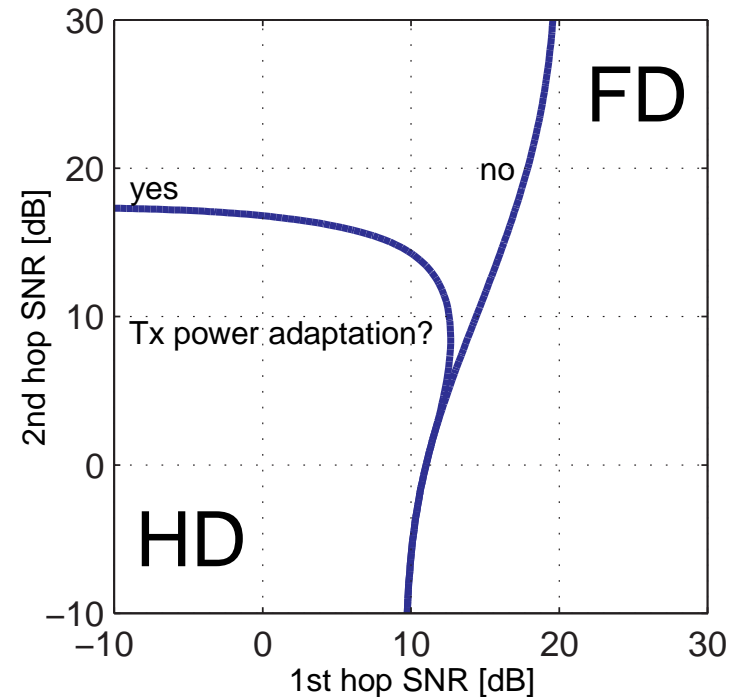


$$\mathcal{R}_{\text{FD}} = \log_2 \left( 1 + \frac{P_S}{P_I + P_N} \right)$$

- We have analyzed various aspects of this tradeoff earlier in the case of two-hop SISO relays [WCNC'09, SPAWC'09, PIMRC'10]

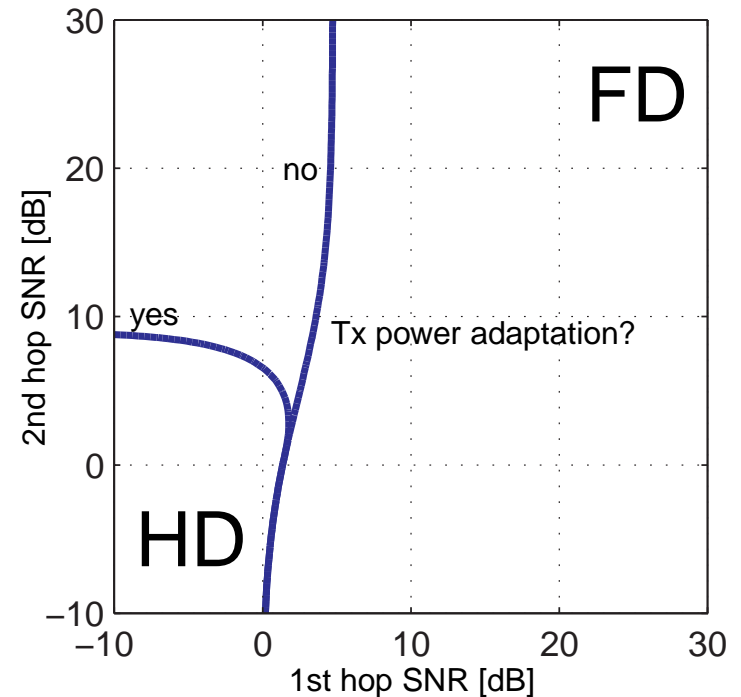
# Motivation from Our SISO Relay Studies (2)

- Half duplex (HD) vs. full duplex (FD)
  - ▷ Which mode gives the best spectral efficiency?
    - More details in [SPAWC'09]
  - ▷ Fig.: Decision boundaries for 10dB interference to noise ratio



# Motivation from Our SISO Relay Studies (2)

- Half duplex (HD) vs. full duplex (FD)
  - ▷ Which mode gives the best spectral efficiency?
    - More details in [SPAWC'09]
  - ▷ Fig.: Decision boundaries for 3dB interference to noise ratio

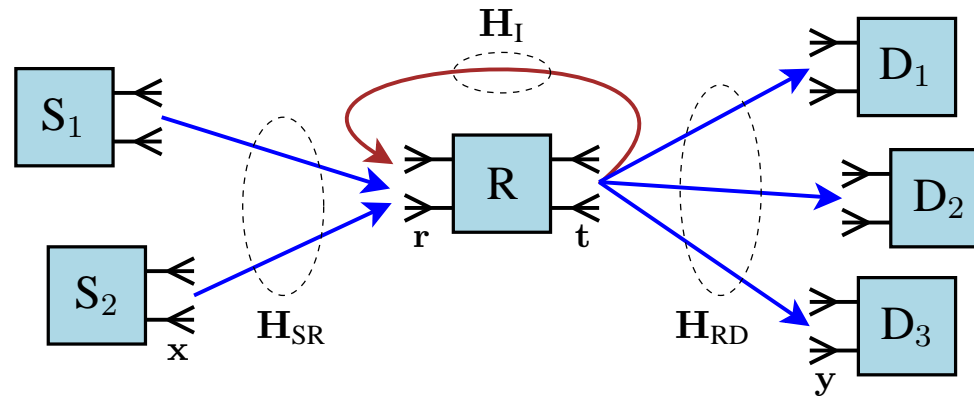


- This tradeoff can be extrapolated to the case of MIMO relays
  - ⇒ Development of mitigation schemes to improve isolation
  - ⇒ Evaluation of self-interference remaining in practice



# System Model

# Signal Model



- Two-hop transmission through a full-duplex MIMO relay:

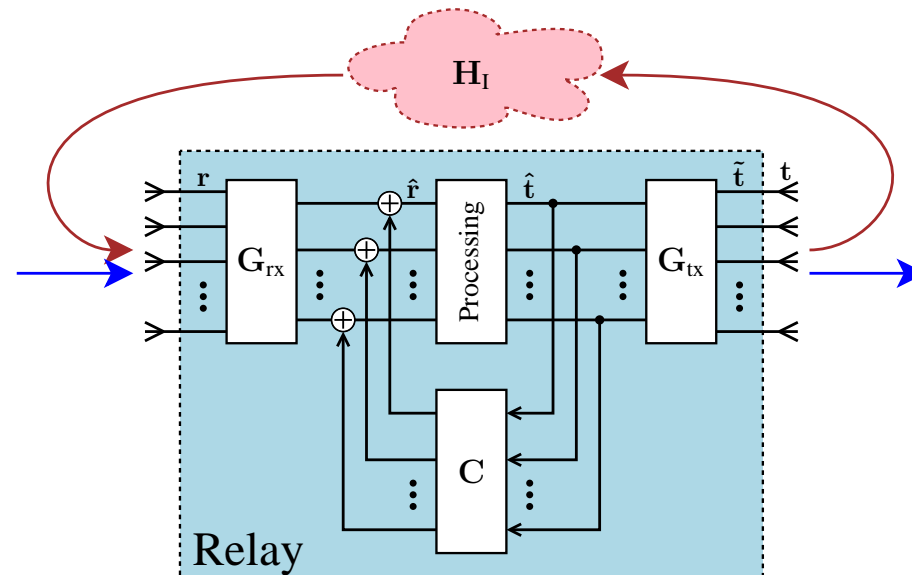
$$\left\{ \begin{array}{l} \text{Sources transmit: } \mathbf{x} \in \mathbb{C}^{N_S \times 1} \\ \text{Relay receives: } \mathbf{r} = \mathbf{H}_{SR}\mathbf{x} + \mathbf{H}_I\mathbf{t} + \mathbf{n}_R \in \mathbb{C}^{N_{rx} \times 1} \\ \text{Relay transmits: } \mathbf{t} \in \mathbb{C}^{N_{tx} \times 1} \\ \text{Destinations receive: } \mathbf{y} = \mathbf{H}_{RD}\mathbf{t} + \mathbf{H}_{SD}\mathbf{x} + \mathbf{n}_D \in \mathbb{C}^{N_D \times 1} \end{array} \right.$$

# Side Information for Transparent Mitigation

- We concentrate on *transparent* mitigation techniques
  - ▷  $\mathbf{t}$  is known by design and  $\mathbf{H}_I$  can be estimated in the relay
  - ▷ Feedback overhead is avoided
- However, the side information can still be imperfect in practice
  - ▷ Mitigation may use only the estimates  $\tilde{\mathbf{H}}_I, \tilde{\mathbf{t}}$  of  $\mathbf{H}_I, \mathbf{t}$ 
    - Channel estimation error:  $\Delta\tilde{\mathbf{H}}_I = \mathbf{H}_I - \tilde{\mathbf{H}}_I$
    - Transmit signal noise:  $\Delta\tilde{\mathbf{t}} = \mathbf{t} - \tilde{\mathbf{t}}$
- The noise in the side information is modeled as i.i.d. Gaussian
  - ▷  $\epsilon_{\mathbf{H}}^2 = \mathcal{E}\{\|\Delta\tilde{\mathbf{H}}_I\|_F^2\} / \mathcal{E}\{\|\mathbf{H}_I\|_F^2\}$
  - ▷  $\epsilon_{\mathbf{t}}^2 = \mathcal{E}\{\|\Delta\tilde{\mathbf{t}}\|_2^2\} / \mathcal{E}\{\|\mathbf{t}\|_2^2\}$

# Mitigation of Self-Interference

# Mitigation of Self-Interference

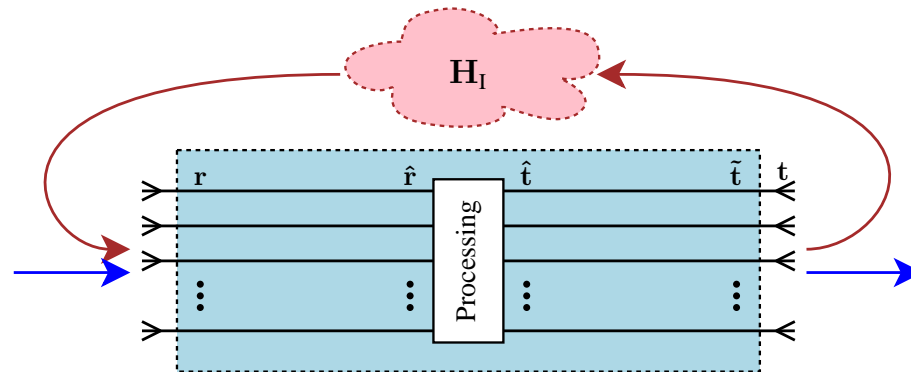


- Cancellation filter  $C$ , spatial suppression filters  $G_{rx}$  and  $G_{tx}$
- Transforms the relay with  $N_{rx} \times N_{tx}$  antennas to an “interference-free” relay with  $\hat{N}_{rx} \times \hat{N}_{tx}$  streams:

$$\begin{cases} \hat{\mathbf{r}} = \mathbf{G}_{rx} \mathbf{r} + \mathbf{C} \hat{\mathbf{t}} \in \mathbb{C}^{\hat{N}_{rx} \times 1} \\ \mathbf{t} = \mathbf{G}_{tx} \hat{\mathbf{t}} + \Delta \tilde{\mathbf{t}} \in \mathbb{C}^{N_{tx} \times 1} \end{cases}$$

- Any protocol can be used for generating  $\hat{\mathbf{t}}$  based on  $\hat{\mathbf{r}}$

# Natural Isolation

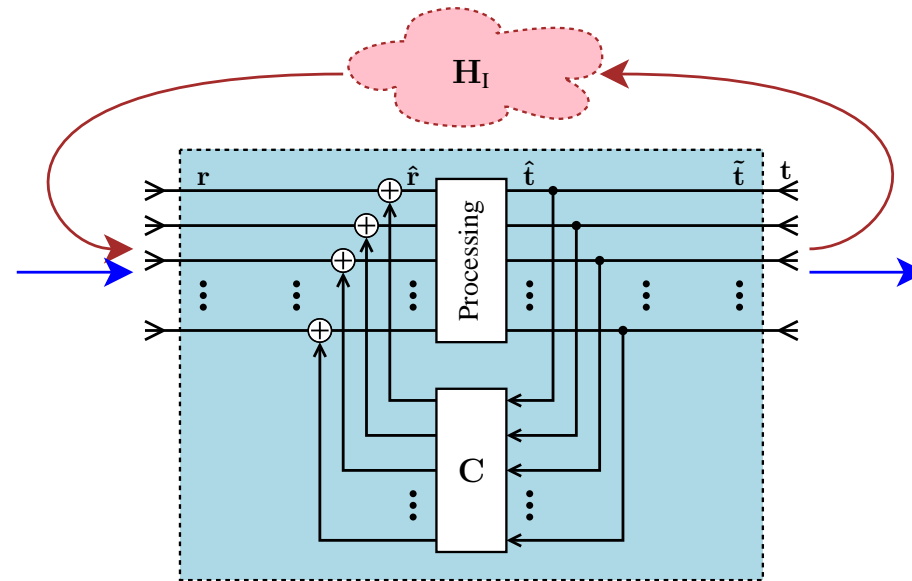


- The reference case without mitigation:  $\mathbf{C} = \mathbf{0}$ ,  $\mathbf{G}_{\text{rx}} = \mathbf{I}$ ,  $\mathbf{G}_{\text{tx}} = \mathbf{I}$
- Self-interference power becomes

$$P_I = \mathcal{E}\{\|\mathbf{H}_I \mathbf{t}\|_2^2\} = (1 + \epsilon_t^2) \text{tr}\{\mathbf{H}_I \mathbf{R}_{\hat{\mathbf{t}}} \mathbf{H}_I^H\} = N_{\text{rx}}(1 + \epsilon_t^2) P_{\text{tx}}$$

- ▶ We assume that  $\mathbf{R}_{\hat{\mathbf{t}}} = \mathcal{E}\{\hat{\mathbf{t}} \hat{\mathbf{t}}^H\} = \frac{P_{\text{tx}}}{N_{\text{tx}}} \mathbf{I}$   
and normalize the channel gain as  $\|\mathbf{H}_I\|_F^2 = N_{\text{rx}} N_{\text{tx}}$

# Time-Domain Cancellation (TDC)

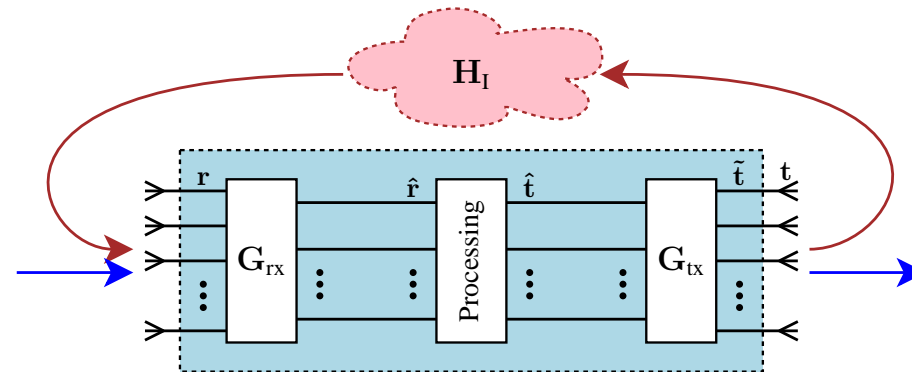


- No spatial filters:  $\mathbf{G}_{\text{rx}} = \mathbf{I}$ ,  $\mathbf{G}_{\text{tx}} = \mathbf{I}$
- Subtract an estimated interference signal from the input signal

$$\mathbf{C} = -\tilde{\mathbf{H}}_{\text{I}}$$

- ▷ Cancellation is imperfect because  $\tilde{\mathbf{H}}_{\text{I}} \neq \mathbf{H}_{\text{I}}$  in practice
- ▷ The unknown transmit signal noise  $\Delta\tilde{\mathbf{t}}$  cannot be cancelled

# Null-Space Projection (NSP)



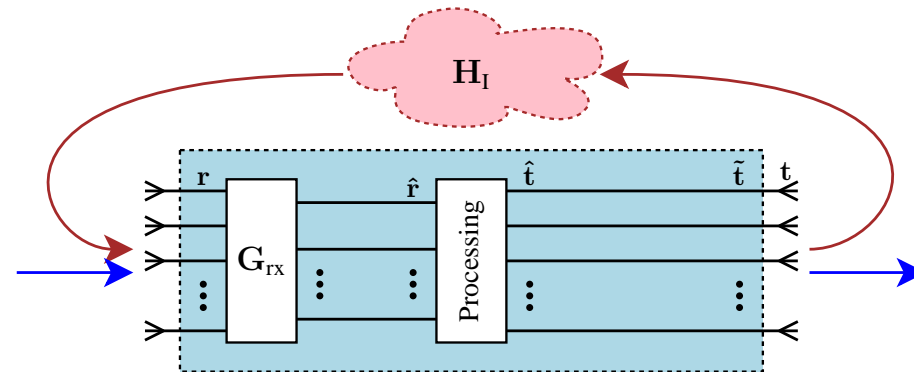
- No cancellation filter:  $\mathbf{C} = \mathbf{0}$
- Choose  $\mathbf{G}_{\text{rx}} = \sqrt{N_{\text{rx}}/\hat{N}_{\text{rx}}} \mathbf{S}_{\text{rx}}^T \tilde{\mathbf{U}}^H$ ,  $\mathbf{G}_{\text{tx}} = \sqrt{N_{\text{tx}}/\hat{N}_{\text{tx}}} \tilde{\mathbf{V}} \mathbf{S}_{\text{tx}}$  such that

$$\mathbf{G}_{\text{rx}} \tilde{\mathbf{H}}_I \mathbf{G}_{\text{tx}} = \mathbf{G}_{\text{rx}} \tilde{\mathbf{U}} \tilde{\Sigma} \tilde{\mathbf{V}}^H \mathbf{G}_{\text{tx}} = \sqrt{\frac{N_{\text{rx}} N_{\text{tx}}}{\hat{N}_{\text{rx}} \hat{N}_{\text{tx}}}} \mathbf{S}_{\text{rx}}^T \tilde{\Sigma} \mathbf{S}_{\text{tx}} = \mathbf{0}$$

- ▷  $\hat{N}_{\text{rx}} + \hat{N}_{\text{tx}} = \max\{N_{\text{rx}}, N_{\text{tx}}\}$  when  $\tilde{\mathbf{H}}_I$  is of full rank
- ▷ Example selection matrices:  $\mathbf{S}_{\text{rx}}^T = [\mathbf{0} | \mathbf{I}]$  and  $\mathbf{S}_{\text{tx}} = [\mathbf{I} | \mathbf{0}]^T$  when  $N_{\text{rx}} \geq N_{\text{tx}}$



# NSP for Suppressing Transmit Signal Noise



- Transmit filter cannot suppress the transmit signal noise:  $\mathbf{G}_{\text{tx}} = \mathbf{I}$
- Choose  $\mathbf{G}_{\text{rx}} = \sqrt{N_{\text{rx}}/\hat{N}_{\text{rx}}} \mathbf{S}_{\text{rx}}^T \tilde{\mathbf{U}}^H$  such that

$$\mathbf{G}_{\text{rx}} \tilde{\mathbf{H}}_{\text{I}} = \mathbf{G}_{\text{rx}} \tilde{\mathbf{U}} \tilde{\Sigma} \tilde{\mathbf{V}}^H = \sqrt{\frac{N_{\text{rx}}}{\hat{N}_{\text{rx}}}} \mathbf{S}_{\text{rx}}^T \tilde{\Sigma} \tilde{\mathbf{V}}^H = \mathbf{0}$$

which also implies  $\mathbf{G}_{\text{rx}} \tilde{\mathbf{H}}_{\text{I}} \mathbf{G}_{\text{tx}} = \mathbf{0}$

- ▷ Unique choice for the row selection matrix:  $\mathbf{S}_{\text{rx}}^T = [\mathbf{0} | \mathbf{I}]$
- ▷  $\hat{N}_{\text{rx}} = N_{\text{rx}} - N_{\text{tx}}$  and  $N_{\text{rx}} > N_{\text{tx}}$  when  $\tilde{\mathbf{H}}_{\text{I}}$  is of full rank

# Residual Self-Interference

# Isolation Improvement

- Additional isolation from mitigation w.r.t. natural isolation:

$$\Delta P_I = \frac{P_I|_{\text{natural}}}{P_I|_{\text{mitigation}}} = \frac{N_{\text{rx}}(1 + \epsilon_{\text{t}}^2)P_{\text{tx}}}{P_I|_{\text{mitigation}}}$$

which is a random variable due to noisy side information

- The residual self-interference can be analyzed in terms of
  - ▶ cumulative distribution function:  $F_{\Delta P_I}(\cdot)$
  - ▶ average improvement of isolation:  $\mathcal{E}\{\Delta P_I\}$

# Some Statistical Tools

- Let  $X$  be a gamma random variable with shape  $k$  and scale  $\theta$ , i.e.,  $X \sim \mathcal{G}(k, \theta)$ 
  - ▷ Cumulative distribution function:  $F_X(x) = 1 - \frac{\Gamma(k, x/\theta)}{\Gamma(k)}$  for  $x \geq 0$
- We will see random variables of the form

$$Y = \frac{a}{X + b}, \quad a \geq 0, b \geq 0$$

- ▷ CDF:  $F_Y(y) = 1 - F_X\left(\frac{a}{y} - b\right)$  for  $0 \leq y \leq \frac{a}{b}$
- ▷ Average:  $\mathcal{E}\{Y\} = \frac{a}{\theta} \exp\left(\frac{b}{\theta}\right) E_k\left(\frac{b}{\theta}\right)$ 
  - $E_k(\cdot)$  is the exponential integral

# Time-Domain Cancellation (TDC)

- Residual self-interference power is

$$P_I = \mathcal{E}\{\|\Delta\tilde{\mathbf{H}}_I\hat{\mathbf{t}} + \mathbf{H}_I\Delta\tilde{\mathbf{t}}\|_2^2\} = N_{\text{rx}} \left( \frac{\|\Delta\tilde{\mathbf{H}}_I\|_F^2}{N_{\text{rx}}N_{\text{tx}}} + \epsilon_{\mathbf{t}}^2 \right) P_{\text{tx}}$$

- Thus, isolation improvement becomes

$$\Delta P_I = \frac{1 + \epsilon_{\mathbf{t}}^2}{X + \epsilon_{\mathbf{t}}^2}$$

where  $X \sim \mathcal{G} \left( N_{\text{rx}}N_{\text{tx}}, \frac{\epsilon_{\mathbf{H}}^2}{N_{\text{rx}}N_{\text{tx}}} \right)$

# Null-Space Projection (NSP)

- Residual self-interference power is

$$\begin{aligned}
 P_I &= \mathcal{E} \{ \|\mathbf{G}_{\text{rx}} \Delta \tilde{\mathbf{H}}_I \mathbf{G}_{\text{tx}} \hat{\mathbf{t}} + \mathbf{G}_{\text{rx}} \mathbf{H}_I \Delta \tilde{\mathbf{t}}\|_2^2 \} \\
 &= N_{\text{rx}} \left( \frac{\|\mathbf{S}_{\text{rx}}^T \tilde{\mathbf{U}}^H \Delta \tilde{\mathbf{H}}_I \tilde{\mathbf{V}} \mathbf{S}_{\text{tx}}\|_F^2}{\hat{N}_{\text{rx}} \hat{N}_{\text{tx}}} + \epsilon_t^2 \frac{\|\mathbf{S}_{\text{rx}}^T \tilde{\mathbf{U}}^H \mathbf{H}_I\|_F^2}{\hat{N}_{\text{rx}} N_{\text{tx}}} \right) P_{\text{tx}}
 \end{aligned}$$

where  $\|\mathbf{S}_{\text{rx}}^T \tilde{\mathbf{U}}^H \mathbf{H}_I\|_F^2 \approx \|\mathbf{S}_{\text{rx}}^T \mathbf{U}^H \mathbf{H}_I\|_F^2$  (a constant)

- Thus, isolation improvement becomes

$$\Delta P_I \approx \frac{1 + \epsilon_t^2}{X + \epsilon_t^2 A}$$

where  $X \sim \mathcal{G} \left( \hat{N}_{\text{rx}} \hat{N}_{\text{tx}}, \frac{\epsilon_{\mathbf{H}}^2}{\hat{N}_{\text{rx}} \hat{N}_{\text{tx}}} \right)$  and  $A = \frac{\|\mathbf{S}_{\text{rx}}^T \mathbf{U}^H \mathbf{H}_I\|_F^2}{\hat{N}_{\text{rx}} N_{\text{tx}}}$

# NSP for Suppressing Transmit Signal Noise

- Residual self-interference power is

$$\begin{aligned} P_I &= \mathcal{E}\{\|\mathbf{G}_{\text{rx}}\Delta\tilde{\mathbf{H}}_I\hat{\mathbf{t}} + \mathbf{G}_{\text{rx}}\Delta\tilde{\mathbf{H}}_I\Delta\tilde{\mathbf{t}}\|_2^2\} \\ &= N_{\text{rx}}(1 + \epsilon_t^2) \frac{\|\mathbf{S}_{\text{rx}}^T \tilde{\mathbf{U}}^H \Delta\tilde{\mathbf{H}}_I\|_F^2}{\hat{N}_{\text{rx}}N_{\text{tx}}} P_{\text{tx}} \end{aligned}$$

- Thus, isolation improvement becomes

$$\Delta P_I = \frac{1}{X}$$

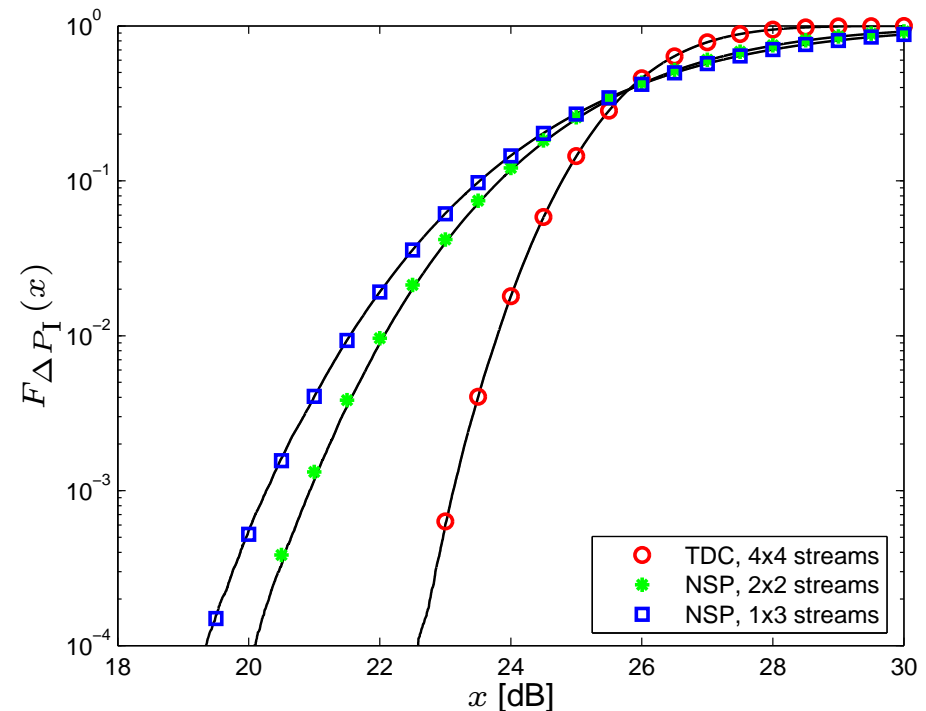
where  $X \sim \mathcal{G}\left(\hat{N}_{\text{rx}}N_{\text{tx}}, \frac{\epsilon_{\mathbf{H}}^2}{\hat{N}_{\text{rx}}N_{\text{tx}}}\right)$

# Numerical Results



# Comparison in Terms of CDF (1)

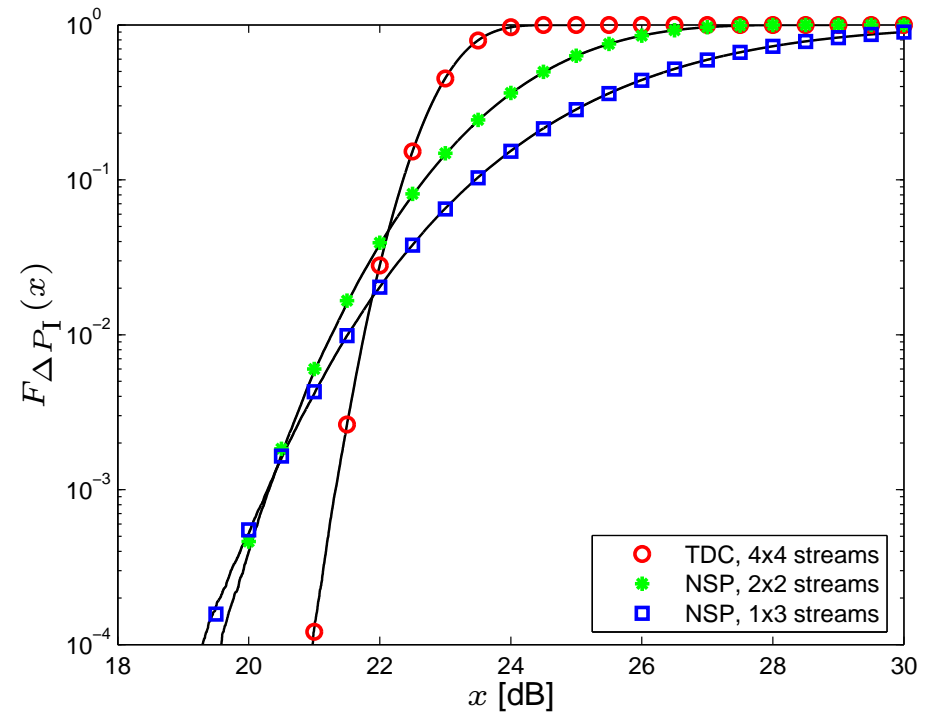
- $N_{\text{rx}} \times N_{\text{tx}} = 4 \times 4$
- $\epsilon_{\mathbf{H}} = 0.05$
- $\epsilon_{\mathbf{t}} = 0$



- TDC: More streams, lower average isolation, lower variance
- NSP: Less streams, higher average isolation, higher variance

# Comparison in Terms of CDF (2)

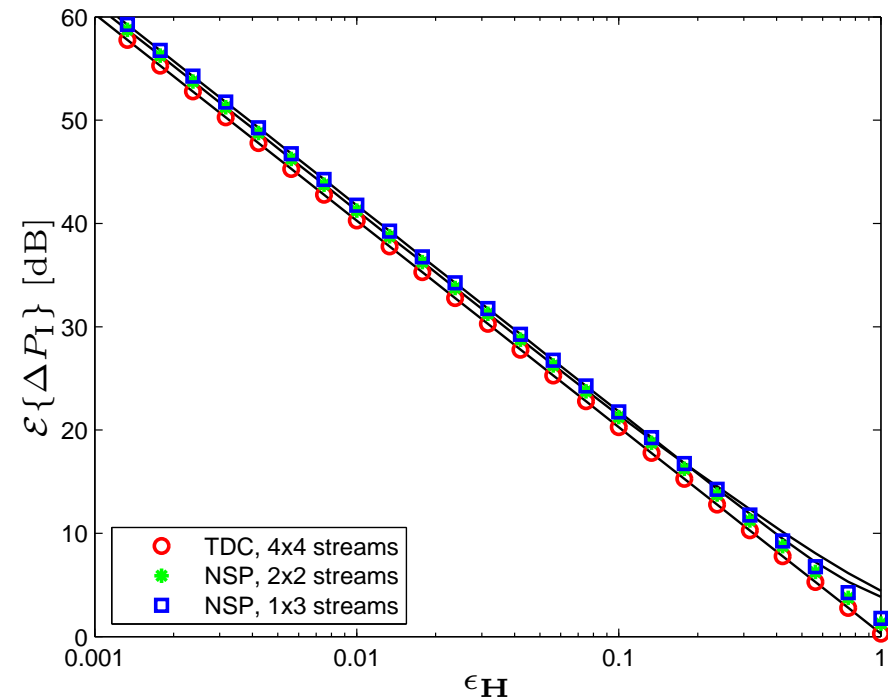
- $N_{\text{rx}} \times N_{\text{tx}} = 4 \times 4$
- $\epsilon_{\mathbf{H}} = 0.05$
- $\epsilon_t = 0.05$



- Transmit signal noise affects more TDC than NSP

# Expected Isolation Improvement (1)

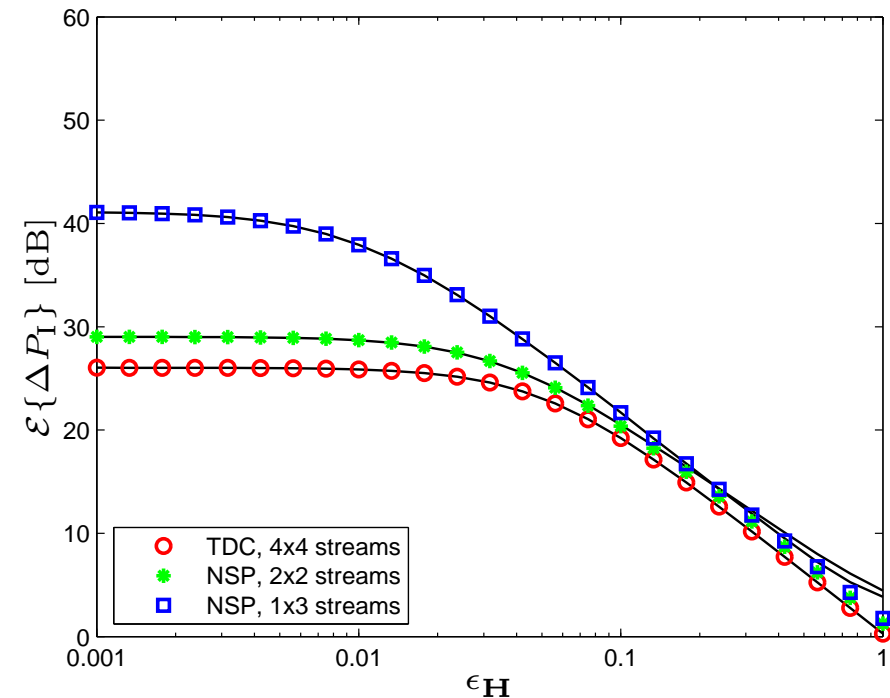
- $N_{\text{rx}} \times N_{\text{tx}} = 4 \times 4$
- $\epsilon_t = 0$



- Isolation improvement decreases linearly as a function of channel estimation error when there is no transmit signal noise

# Expected Isolation Improvement (2)

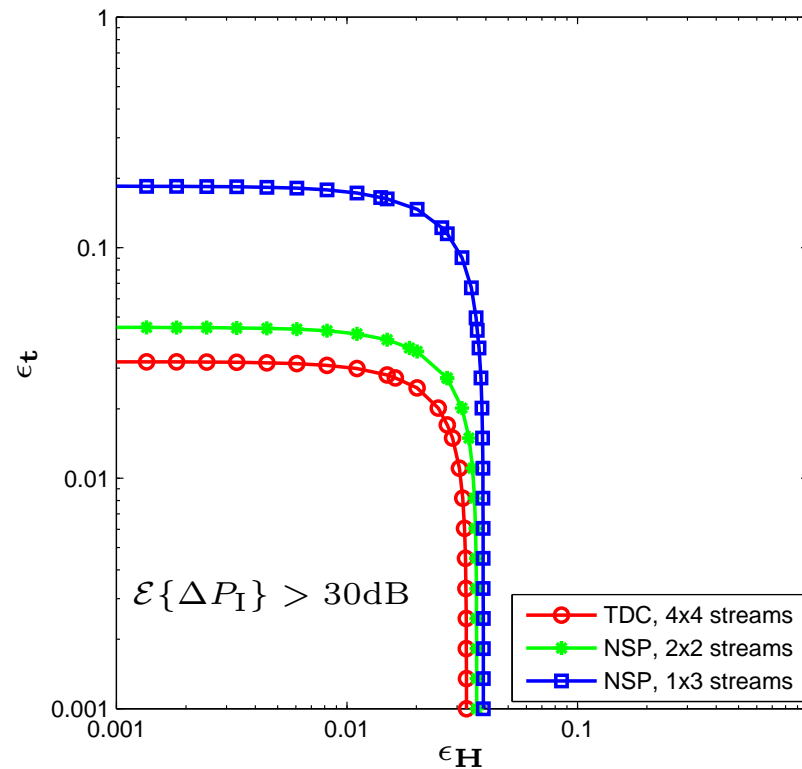
- $N_{\text{rx}} \times N_{\text{tx}} = 4 \times 4$
- $\epsilon_t = 0.05$



- Transmit signal noise limits the maximum isolation improvement

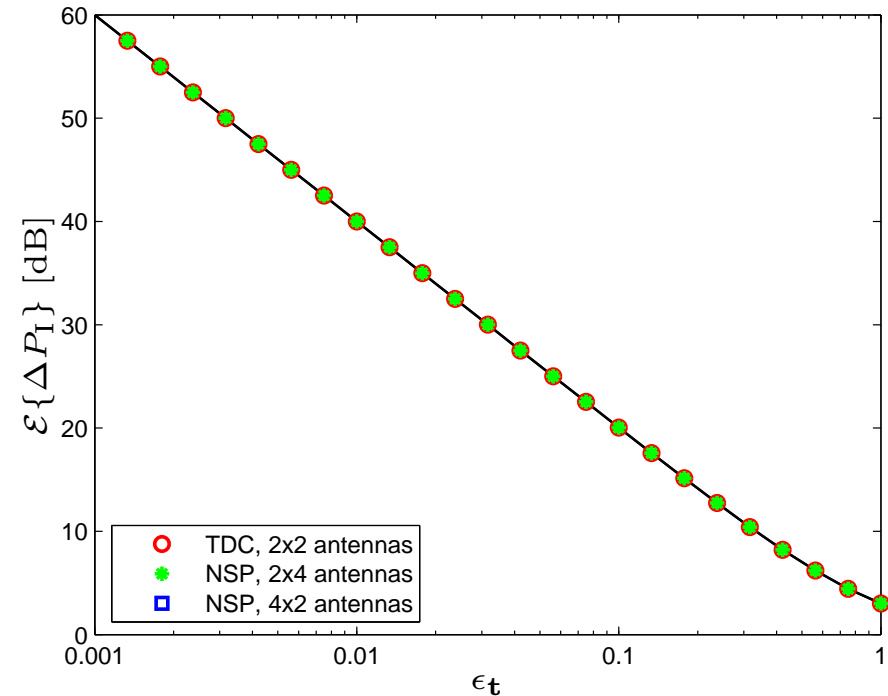
# Robustness to Noisy Side Information

- When is isolation improved by at least 30dB?
- $N_{\text{rx}} \times N_{\text{tx}} = 4 \times 4$



# NSP for Suppressing Transmit Signal Noise (1)

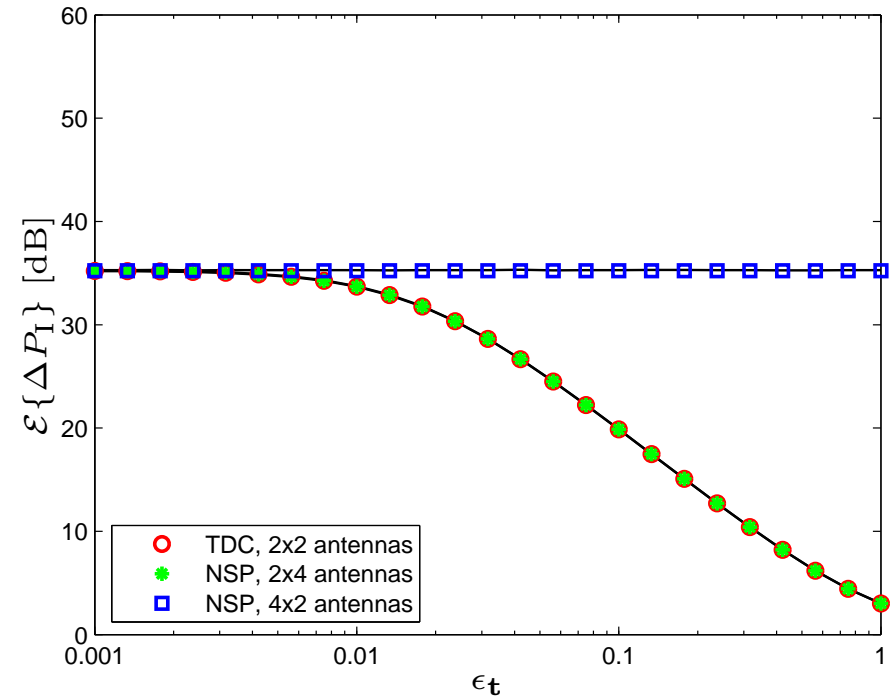
- $\hat{N}_{\text{rx}} \times \hat{N}_{\text{tx}} = 2 \times 2$
- $\epsilon_{\mathbf{H}} = 0$



- Isolation improvement decreases linearly as a function of transmit signal noise level when there is no channel estimation error

# NSP for Suppressing Transmit Signal Noise (2)

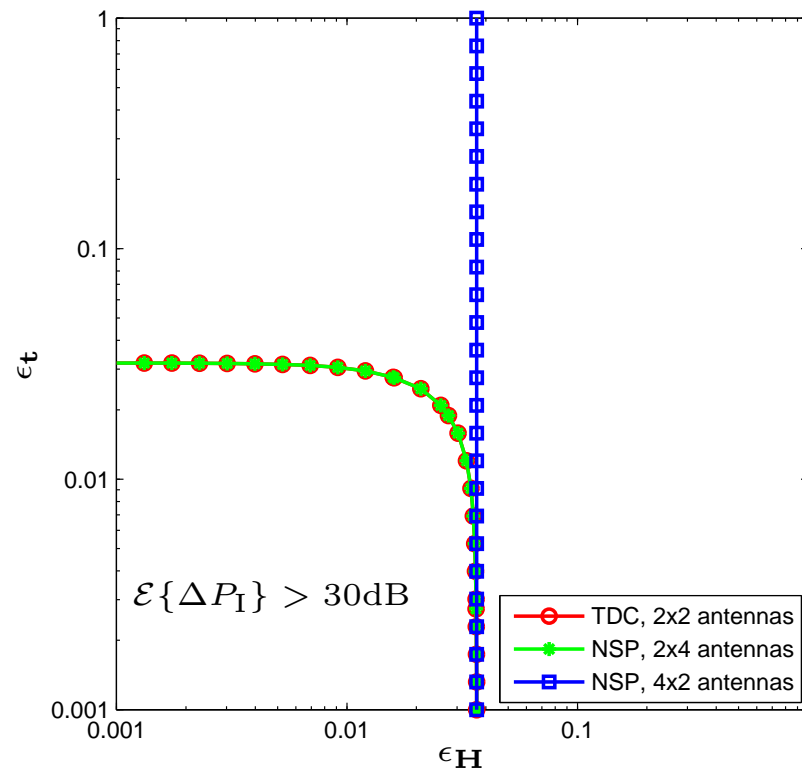
- $\hat{N}_{\text{rx}} \times \hat{N}_{\text{tx}} = 2 \times 2$
- $\epsilon_{\text{H}} = 0.02$



- NSP can be made immune to transmit signal noise

# Robustness to Noisy Side Information

- When is isolation improved by at least 30dB?
- $\hat{N}_{\text{rx}} \times \hat{N}_{\text{tx}} = 2 \times 2$





# Conclusion

# Conclusion

- Full-duplex relaying has potential to improve spectral efficiency w.r.t. conventional half-duplex relaying
- Main technical problem: self-interference at the relay
  - ▷ Separated antenna arrays in the relay: natural isolation
  - ▷ Signal processing techniques to improve isolation further
    - Time-domain cancellation
    - Spatial-domain suppression: Null-space projection
- Analytical evaluation of self-interference remaining in practice
  - ▷ Side information used in mitigation cannot be perfect
  - ▷ Additional isolation given by mitigation schemes is specified by the noise levels of the side information



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